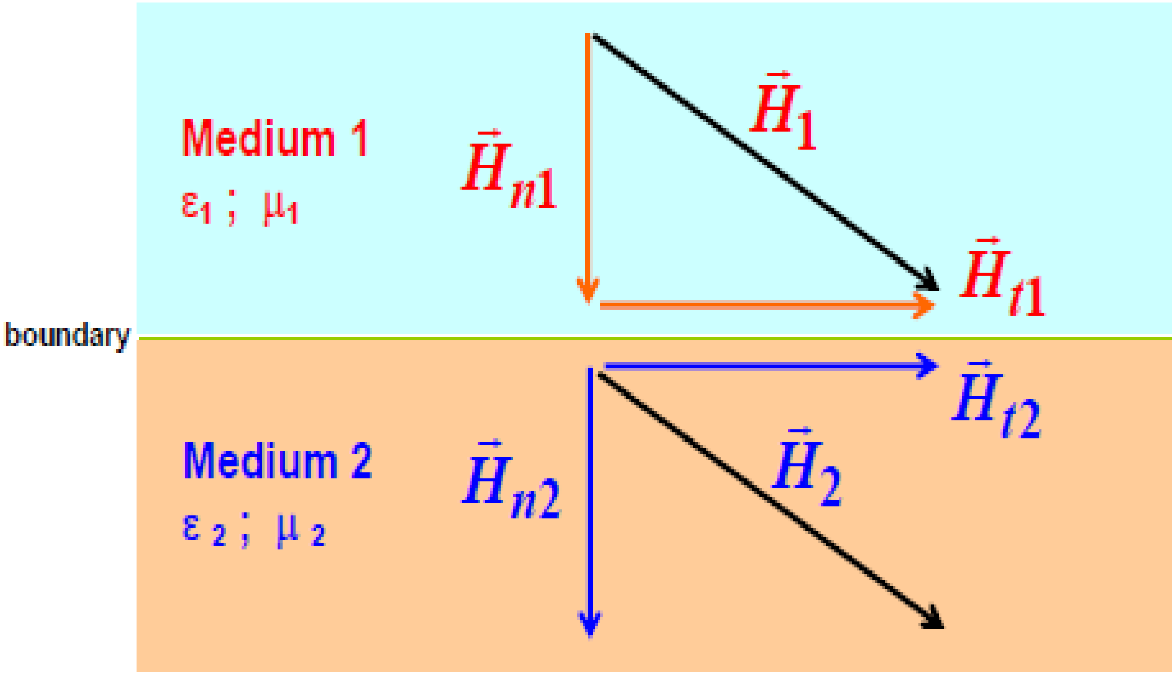
Review of Boundary Conditions

Consider an electromagnetic field at the boundary between two materials with different properties. The tangent and the normal component of the fields must me examined separately, in order to understand the effects of the boundary.



Summary

$$\boxed{\hat{n} \times (\vec{\mathrm{H}}_{t1} - \vec{\mathrm{H}}_{t2}) = \vec{J}_s}$$

$$\boxed{\hat{n} \times (\vec{\mathrm{E}}_{t1} - \vec{\mathrm{E}}_{t2}) = 0}$$

$$\boxed{\mathrm{D}_{1n} - \mathrm{D}_{2n} = \rho_s \qquad \mathrm{B}_{1n} - \mathrm{B}_{2n} = 0}$$

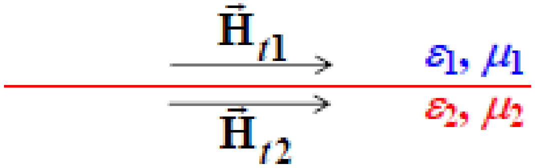
$\hat{n}$ = unit vector normal to the surface

n is a unit vector pointing from region 2 into region 1

Conditions on E and H

SUMMARY

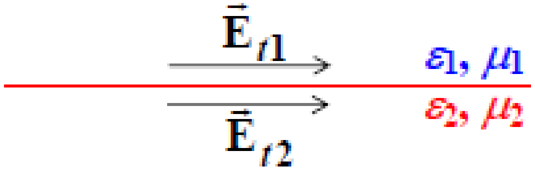
If medium 2 is perfect conductor



$$\vec{H}_{t1} = \vec{H}_{t2}$$

$$\hat{n} \times \vec{H}_{t1} = \vec{J}_s$$

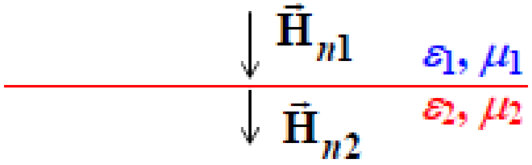
$$\vec{H}_{t2} = 0$$



$$\vec{E}_{t1} = \vec{E}_{t2}$$

$$\vec{E}_{t1} = 0$$

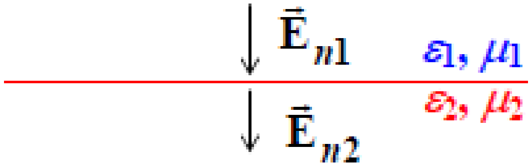
$$\vec{E}_{t2} = 0$$



$$\mu_1 \vec{H}_{n1} = \mu_2 \vec{H}_{n2}$$

$$\vec{H}_{n1} = 0$$

$$\vec{H}_{n2} = 0$$



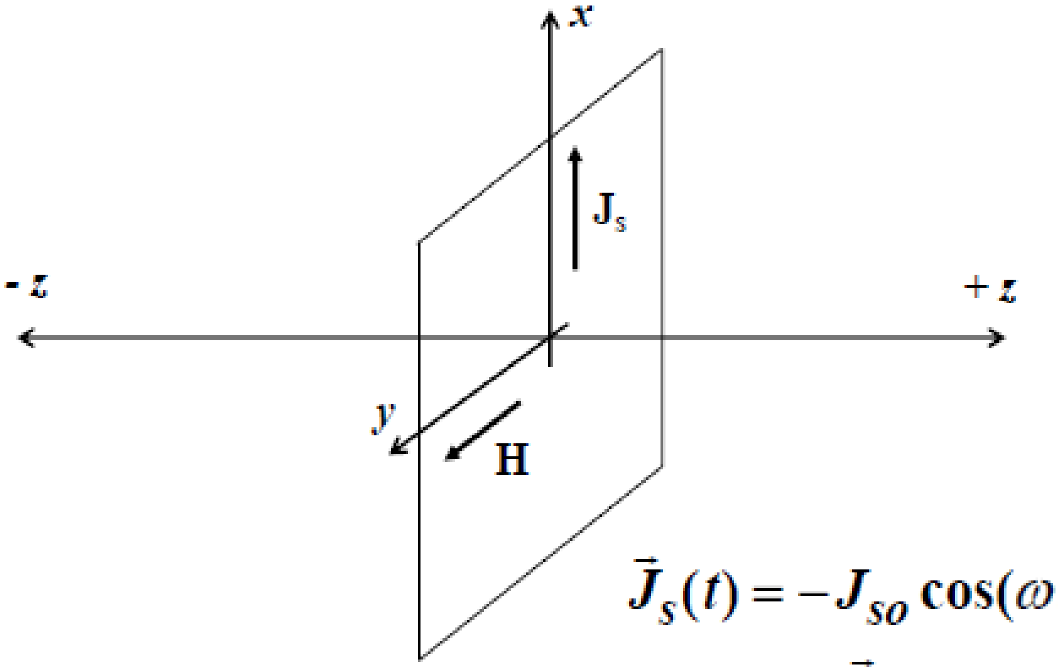
$$\varepsilon_1 \vec{E}_{n1} = \varepsilon_2 \vec{E}_{n2} + \rho_s$$

$$\vec{E}_{n1} = \rho_s / \varepsilon_1$$

$$\vec{E}_{n2} = 0$$

Example

An infinite current sheet generates a plane wave (free space on both sides)



$$\vec{J}_s(t) = -J_{so}\cos(\omega t)\,\hat{i}_x$$

$$\text{Phasor}\quad \vec{\mathbf{J}}_s = -J_{so}\,\hat{i}_x$$

The E.M. field is transmitted on both sides of the infinitesimally thin sheet of current.

Solution

BOUNDARY CONDITIONS

$$\boxed{\hat{n}\times(\vec{\mathrm{H}}_{t1}-\vec{\mathrm{H}}_{t2})=\vec{\mathrm{J}}_s}$$

$$\vec{\mathrm{H}}_{t1}-\vec{\mathrm{H}}_{t2}=J_{so}\hat{i}_x$$

$$\vec{\mathrm{E}}_{t1}=\vec{\mathrm{E}}_{t2}$$

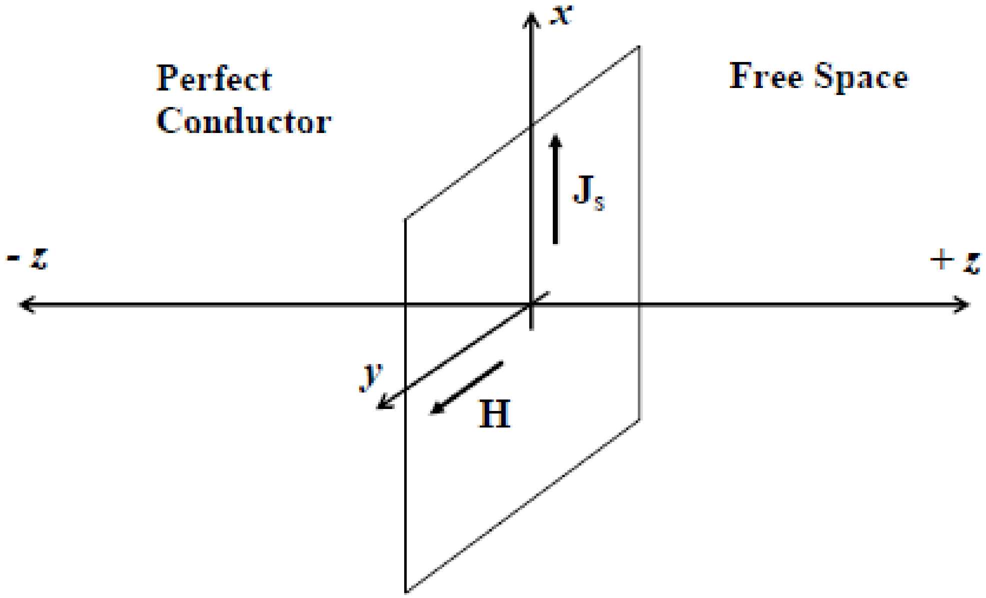
$$\left|\vec{\mathrm{E}}_{t1}\right|=\eta_0\left|\vec{\mathrm{H}}_{t1}\right|$$

$$\text{Symmetry}\ \Rightarrow\left|\vec{\mathrm{H}}_{t1}\right|=\left|\vec{\mathrm{H}}_{t2}\right|$$

$$\mathrm{H}_1=\frac{J_{so}}{2}\qquad\qquad \mathrm{H}_2=-\frac{J_{so}}{2}$$

Example

A semi-infinite perfect conductor medium in contact with free space has uniform surface current and generates a plane wave



$$\vec{\mathbf{J}}_s = -J_{so}\cos(\omega t)\,\hat{i}_x$$

The E.M. field is zero inside the perfect conductor. The wave is only transmitted into free space.

Solution

BOUNDARY CONDITIONS

$$\boxed{\hat{n}\times(\vec{\mathbf{H}}_{t1}-\vec{\mathbf{H}}_{t2})=\vec{\mathbf{J}}_s}$$

$$\vec{\mathbf{H}}_{t1}-\vec{\mathbf{H}}_{t2}=\vec{\mathbf{H}}_{t1}-0=J_{so}\,\hat{i}_x$$

$$\vec{\mathbf{E}}_{t2}=0$$

$$\text{Asymmetry}\ \Rightarrow\left|\vec{\mathbf{H}}_{t1}\right|\neq\left|\vec{\mathbf{H}}_{t2}\right|$$

$$\mathbf{H}_{t1}=J_{so}\qquad\qquad \mathbf{H}_{t2}=0$$

Electromagnetic Waves in Material Media

In a material medium free charges may be present, which generate a current under the influence of the wave electric field. The current J_c is related to the electric field E through the conductivity σ as

$$\boxed{\vec{J}_c = \sigma \vec{E}}$$

The material may also have specific relative values of dielectric permittivity and magnetic permeability

$$\boxed{\varepsilon = \varepsilon_r \varepsilon_o \qquad \mu = \mu_r \mu_o}$$

Maxwell's equations

$$\nabla \times \vec{\mathbf{E}} = -j\omega\mu\,\vec{\mathbf{H}}$$

$$\nabla \times \vec{\mathbf{H}} = \sigma\,\vec{\mathbf{E}} + j\omega\varepsilon\,\vec{\mathbf{E}} = j\omega(\varepsilon - j\frac{\sigma}{\omega})\vec{\mathbf{E}}$$

In phasor notation, it is as if the material conductivity introduces an imaginary part for the dielectric constant ε. The wave equation for the phasor electric field is given by

$$\nabla \times \nabla \times \vec{\mathbf{E}} = \nabla\nabla \cdot \vec{\mathbf{E}} - \nabla^2\vec{\mathbf{E}} = -j\omega\mu\,\nabla \times \vec{\mathbf{H}}$$

$$= -j\omega\mu(\vec{\mathbf{J}}_{\mathbf{c}} + j\omega\varepsilon\,\vec{\mathbf{E}})$$

$$\Rightarrow \quad \nabla^2\vec{\mathbf{E}} = j\omega\mu(\sigma + j\omega\varepsilon)\vec{\mathbf{E}}$$

We have assumed that the net charge density is zero, even if a conductivity is present, so that the electric field divergence is zero.

In 1-D the wave equation is

$$\frac{\partial^2 \vec{\mathrm{E}}_x}{\partial z^2} = j\omega\mu(\sigma + j\omega\varepsilon)\vec{\mathrm{E}}_x = \gamma^2 \vec{\mathrm{E}}_x$$

with general solution

$$\mathrm{E}_x(z) = \overline{A}\exp(-\gamma z) + \overline{B}\exp(\gamma z)$$

$$H_y(z) = -\frac{1}{j\omega\mu}\frac{\partial E_x}{\partial z} = \sqrt{\frac{\sigma + j\omega\varepsilon}{j\omega\mu}}\left(\overline{A}\exp(-\gamma z) - \overline{B}\exp(\gamma z)\right)$$

$$= \frac{1}{\overline{\eta}}\left(\overline{A}\exp(-\gamma z) - \overline{B}\exp(\gamma z)\right)$$

These resemble the voltage and current solutions in lossy transmission lines.

Characteristic Impedance and Propagation Constant

The intrinsic impedance of the medium is defined as

$$\bar{\eta} = |\bar{\eta}|\, e^{j\tau} = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\varepsilon}}$$

For the propagation constant, one can obtain the real and imaginary parts as

$$\gamma = \sqrt{j\omega\mu(\sigma + j\omega\varepsilon)} = \alpha + j\beta$$

$$\alpha = \frac{\omega\sqrt{\mu\varepsilon}}{\sqrt{2}}\left[\sqrt{1+\left(\frac{\sigma}{\omega\varepsilon}\right)^2} - 1\right]^{1/2}$$

$$\beta = \frac{\omega\sqrt{\mu\varepsilon}}{\sqrt{2}}\left[\sqrt{1+\left(\frac{\sigma}{\omega\varepsilon}\right)^2} + 1\right]^{1/2}$$

Phase velocity and wavelength

$$v_p = \frac{\omega}{\beta} = \frac{\sqrt{2}}{\sqrt{\mu\varepsilon}}\left[\sqrt{1+\left(\frac{\sigma}{\omega\varepsilon}\right)^2}+1\right]^{-1/2}$$

$$\lambda = \frac{2\pi}{\beta} = \frac{\sqrt{2}}{f\sqrt{\mu\varepsilon}}\left[\sqrt{1+\left(\frac{\sigma}{\omega\varepsilon}\right)^2}+1\right]^{-1/2}$$

The intrinsic impedance of the medium is complex as long as the conductivity is not zero. The phase angle of the intrinsic impedance indicates that electric field and magnetic field are out of phase. Considering only the forward wave solutions

$$\mathbf{E}_x(z) = \overline{A}\exp(-\gamma\, z) = \overline{A}\exp(-\alpha\, z)\exp(-j\beta\, z)$$

$$\mathbf{H}_y(z) = \frac{1}{|\eta|}\overline{A}\exp(-\gamma\, z - j\tau) = \frac{1}{|\eta|}\overline{A}\exp(-\alpha\, z)\exp(-j\beta\, z - j\tau)$$

In Time-Domain

In time-dependent form

$$E_x(z,t) = \mathrm{Re}\left\{ \left|\overline{A}\right| \exp(j\theta)\exp(-\alpha\, z)\exp(-j\beta\, z)\exp(j\omega\, t)\right\}$$

$$= \left|\overline{A}\right| \exp(-\alpha\, z)\cos(\omega\, t - \beta\, z + \theta)$$

$$H_y(z,t) = \frac{1}{\left|\overline{\eta}\right|}\mathrm{Re}\left\{ \left|\overline{A}\right| \exp(j\theta)\exp(-\alpha\, z)\exp(-j\beta\, z - j\tau)\exp(j\omega\, t)\right\}$$

$$= \frac{1}{\left|\overline{\eta}\right|}\left|\overline{A}\right| \exp(-\alpha\, z)\cos(\omega\, t - \beta\, z + \theta - \tau)$$

where the integration constant has been assumed to be in general a complex quantity as

$$\overline{A} = \left|\overline{A}\right| \exp(j\theta)$$

Classification of materials

Perfect dielectrics - For these materials $\sigma = 0$

Propagation constant

$$\beta = \omega\sqrt{\varepsilon_r \varepsilon_o \mu_r \mu_o}$$

$$\alpha = 0$$

Phase velocity

$$v_p = \frac{\omega}{\beta} = \frac{1}{\sqrt{\mu_r \mu_o \varepsilon_r \varepsilon_o}}$$

Medium Impedance

$$\eta = \sqrt{\frac{j\omega\mu}{j\omega\varepsilon}} = \sqrt{\frac{\mu_r \mu_o}{\varepsilon_r \varepsilon_o}}$$

Wavelength

$$\lambda = \frac{2\pi}{\beta} = \frac{v_p}{f} = \frac{1}{f\sqrt{\mu_r \mu_o \varepsilon_r \varepsilon_o}}$$

Imperfect dielectrics

Imperfect dielectrics – For these materials $\sigma \neq 0$ but $(\sigma/\omega\varepsilon)<<1$

$$\gamma = \sqrt{j\omega\mu(\sigma + j\omega\varepsilon)} = j\omega\sqrt{\mu\varepsilon}\sqrt{1 - j\frac{\sigma}{\omega\varepsilon}}$$

$$\approx \frac{\sigma}{2}\sqrt{\frac{\mu}{\varepsilon}} + j\omega\sqrt{\mu\varepsilon} + \ldots$$

$$\alpha \approx \frac{\sigma}{2}\sqrt{\frac{\mu}{\varepsilon}} \qquad \beta \approx \omega\sqrt{\mu\varepsilon}$$

$$v_p = \frac{\omega}{\beta} \approx \frac{1}{\sqrt{\mu\varepsilon}} \qquad \lambda = \frac{2\pi}{\beta} \approx \frac{1}{f\sqrt{\mu\varepsilon}}$$

$$\eta = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\varepsilon}} = \sqrt{\frac{j\omega\mu}{j\omega\varepsilon}}\left(1 - j\frac{\sigma}{\omega\varepsilon}\right)^{-1/2} \approx \sqrt{\frac{\mu}{\varepsilon}}$$

Good Conductors

Good conductors – For these materials $\sigma \neq 0$ but $(\sigma/\omega\varepsilon) >> 1$

$$\gamma = \sqrt{j\omega\mu(\sigma + j\omega\varepsilon)} \approx \sqrt{j\omega\mu\sigma} = \sqrt{\omega\mu\sigma}\sqrt{j}$$

$$= \sqrt{\omega\mu\sigma}\exp(j\frac{\pi}{4}) = \sqrt{\omega\mu\sigma}\left(\frac{1}{\sqrt{2}} + j\frac{1}{\sqrt{2}}\right) = \sqrt{\pi f\mu\sigma}(1 + j)$$

$$\alpha \approx \sqrt{\pi f\mu\sigma} \qquad \beta \approx \sqrt{\pi f\mu\sigma}$$

$$v_p = \frac{\omega}{\beta} \approx \sqrt{\frac{4\pi f}{\mu\sigma}} \qquad \lambda = \frac{2\pi}{\beta} \approx \sqrt{\frac{4\pi}{f\mu\sigma}}$$

$$\eta = \sqrt{\frac{j\omega\mu}{\sigma + j\omega\varepsilon}} \approx \sqrt{\frac{j\omega\mu}{\sigma}} = \sqrt{\frac{\omega\mu}{\sigma}}\exp(j\frac{\pi}{4})$$

$$= \sqrt{\frac{\omega\mu}{\sigma}}\left(\frac{1}{\sqrt{2}} + j\frac{1}{\sqrt{2}}\right) = \sqrt{\frac{\pi f\mu}{\sigma}}(1 + j)$$

Cont.

The simple rule of thumb is that approximations for good conductor can be applied when

$$\frac{\sigma}{\omega\varepsilon} \geq 10$$

Note that for a good conductor the attenuation constant α and the propagation constant β are approximately equal.

The medium impedance η has nearly equal real and imaginary parts, therefore its phase angle is approximately 45°.

This means that in a good conductor the electric and magnetic fields have always a phase difference $\tau = 45^\circ = \pi/4$.

Cont.

Also, in a good conductor the fields attenuate very rapidly. The distance over which fields are attenuated by a factor exp(–1.0) is

$$\frac{1}{\alpha} = \delta = \frac{1}{\sqrt{\pi f \mu \sigma}} = \text{Skin depth}$$

A typical good conductor is copper, which has the following parameters:

$$\sigma = 5.80 \times 10^{7}\,[\mathrm{S/m}]$$

$$\varepsilon \approx \varepsilon_o$$

$$\mu \approx \mu_o$$

Copper and Sea Water

Copper remains a good conductor at extremely high frequencies. Another good conductor example is sea water at relatively low frequencies

$$\sigma \approx 4.0\ [\mathrm{S/m}]$$

$$\varepsilon \approx 80\varepsilon_o$$

$$\mu \approx \mu_o$$

At a frequency of 25 kHz

$$\frac{\sigma}{\omega\varepsilon} \approx 36,000$$

Perfect Conductor

Perfect conductor - For this ideal material $\sigma \to \infty$

For this material, the attenuation is also infinite and the skin depth goes to zero. This means that the electromagnetic field must go to zero below the perfect conductor surface.

General medium - When a material is not covered by one of the limit cases, the complete formulation must be used. We can classify a material for which the conditions $(\sigma/\omega\varepsilon) \ll 1$ or $(\sigma/\omega\varepsilon) \gg 10$ are invalid as a general medium.

The simple rule of thumb for general medium is

$$10 > \frac{\sigma}{\omega\varepsilon} > 0.1$$